

Motivation

Wireless networks with low-latency requirements are finding numerous applications; such as machine-to-machine (M2M) communications.

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Finite blocklength results inherently depend upon the CDF of the channel mutual information random variable [1] and its statistics, specially the second order statistic known as the channel dispersion [2,3].

Modified Mutual Information

 $i(x^n; y^n) = \log \frac{P_{Y^n | X^n}(y^n | x^n)}{P_{Y^n}(y^n)}$

with non-i.i.d. input: $i(X^n; Y^n) \neq \sum_{t=1}^n i(X_t; Y_t)$ Fix $Q_{Y^n} = \prod_{t=1}^n Q_{Y_t}$ and define the modified mutual information RV

 $\tilde{i}(x^n; y^n) = \log \frac{P_{Y^n | X^n}(y^n | x^n)}{Q_{V^n}(y^n)}$ Then $\tilde{i}(X^n;Y^n) = \sum_{t=1}^n \tilde{i}(X_t;Y_t)$

For Achievability: use modified typicality $\tilde{i}(X^n; Y^n) > \log \gamma_n$ Then, the relevant outage probability is

 $P_{X^n} P_{Y^n | X^n} [\tilde{i}(X^n; Y^n) \le \log \gamma_n]$

Moreover, a change of measure & uniform bound technique takes care of the non-i.i.d. output distribution.

$$P_{X^{n}}P_{Y^{n}}[\tilde{i}(X^{n};Y^{n}) > \log \gamma_{n}]$$

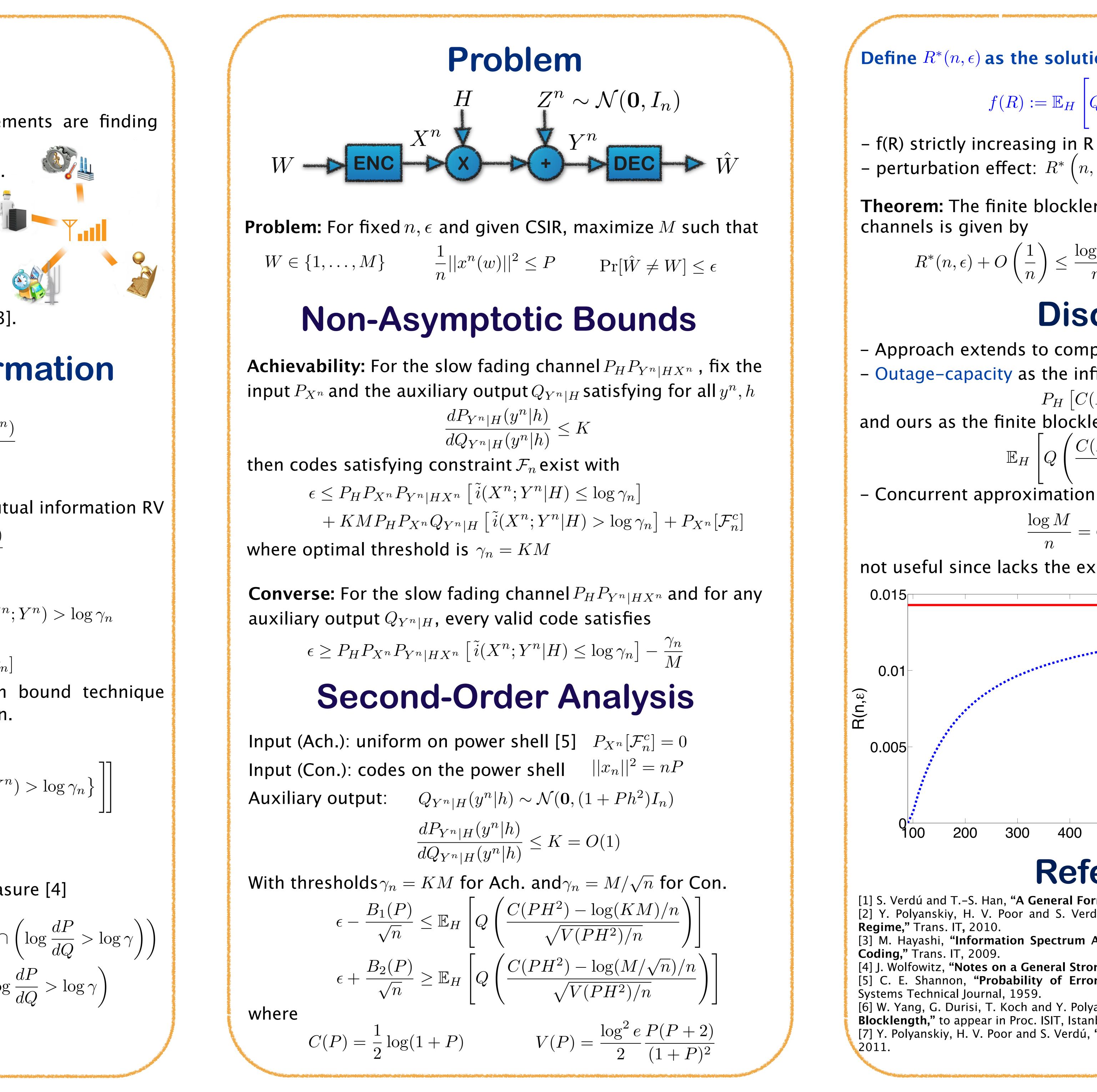
$$= \mathbb{E}_{P_{X^{n}}}\left[\mathbb{E}_{Q_{Y^{n}}}\left[\underbrace{\frac{dP_{Y^{n}}(Y^{n})}{dQ_{Y^{n}}(Y^{n})}}_{\leq K, \forall y^{n}} 1\left\{\tilde{i}(X^{n};Y^{n}) > \underbrace{\frac{dP_{Y^{n}}(Y^{n})}{i(X^{n};Y^{n})}}_{\leq K,\forall y^{n}}\right\}$$

For Outer Bound: use another change of measure [4]

$$\begin{split} P(D) &= P\left(D \cap \left(\log \frac{dP}{dQ} \le \log \gamma\right)\right) + P\left(D \cap \left(\log \frac{dP}{dQ} \le \log \gamma\right)\right) + P\left(\log \frac{dP}{dQ} \le \gamma Q(D) + P\left(\log \frac{dP}{dQ} \le \log \gamma\right)\right) + P\left(\log \frac{dP}{dQ} \le \gamma Q(D) + P\left(\tilde{i} > \log \gamma\right)\right) \end{split}$$



On the Dispersion of Slow Rayleigh Fading Channels Ebrahim MolavianJazi and J. Nicholas Laneman



Define $R^*(n, \epsilon)$ **as the solution to** $f(R) = \epsilon$ where

$$f(R) := \mathbb{E}_H \left[Q \left(\frac{C(PH^2) - R}{\sqrt{V(PH^2)/n}} \right) \right]$$

- perturbation effect: $R^*\left(n, \epsilon + O\left(\frac{1}{\sqrt{n}}\right)\right) = R^*(n, \epsilon) + O\left(\frac{1}{n}\right)$

Theorem: The finite blocklength coding rate of slow fading

$$O\left(\frac{1}{n}\right) \le \frac{\log M}{n} \le R^*(n,\epsilon) + \frac{1}{2}\frac{\log n}{n} + O\left(\frac{1}{n}\right)$$

Discussion

- Approach extends to complex noise and general fading - Outage-capacity as the infinite blocklength performance

$$P_{H} \left[C(PH^{2}) < C_{out} \right] = \epsilon$$

finite blocklength behavior
$$\mathbb{E}_{H} \left[Q \left(\frac{C(PH^{2}) - R^{*}(n, \epsilon)}{\sqrt{V(PH^{2})/n}} \right) \right] =$$

- Concurrent approximation of Wang et al. [6]

$$\frac{\log M}{n} = C_{\text{out}} + O\left(\frac{\log n}{n}\right)$$

not useful since lacks the exact coefficient of the 'log(n)' term

- Outage Cap. 2nd Order Apprx. 300 900 1000 500 600 Blocklength n References [1] S. Verdú and T.-S. Han, "A General Formula for Channel Capacity," Trans. IT, 1994. [2] Y. Polyanskiy, H. V. Poor and S. Verdú, "Channel Coding Rate in the Finite Blocklength [3] M. Hayashi, "Information Spectrum Approach to Second-Order Coding Rate in Channel [4] J. Wolfowitz, "Notes on a General Strong Converse," Information and Control, 1968. [5] C. E. Shannon, "Probability of Error for Optimal Codes in a Gaussian Channel," Bell [6] W. Yang, G. Durisi, T. Koch and Y. Polyanskiy, "Quasi-Static SIMO Fading Channels at Finite **Blocklength,**" to appear in Proc. ISIT, Istanbul, Turkey, July 2013. [7] Y. Polyanskiy, H. V. Poor and S. Verdú, "Dispersion of the Gilbert-Elliott Channel," Trans. IT,

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